

dBCOOL™ Remote Thermal Controller and Fan Controller

ADT7460*

FEATURES

Controls and Monitors up to 4 Fan Speeds 1 On-Chip and 2 Remote Temperature Sensors Dynamic T_{MIN} Control Mode Optimizes System Acoustics Intelligently Automatic Fan Speed Control Mode Controls System Cooling Based on Measured Temperature Enhanced Acoustic Mode Dramatically Reduces User Perception of Changing Fan Speeds Thermal Protection Feature via THERM Output Monitors Performance Impact of Intel® Pentium® 4 Processor Thermal Control Circuit via THERM Input 2-Wire and 3-Wire Fan Speed Measurement Limit Comparison of All Monitored Values Meets SMBus 2.0 Electrical Specifications (Fully SMBus 1.1 Compliant)

APPLICATIONS

Low Acoustic Noise PCs Networking and Telecommunications Equipment

GENERAL DESCRIPTION

The ADT7460 *dB*COOL controller is a thermal monitor and multiple PWM fan controller for noise-sensitive applications requiring active system cooling. It can monitor the temperature of up to two remote sensor diodes, plus its own internal temperature. It can measure and control the speed of up to four fans so that they operate at the lowest possible speed for minimum acoustic noise. The Automatic Fan Speed Control Loop optimizes fan speed for a given temperature. A unique Dynamic T_{MIN} Control Mode enables the system thermals/acoustics to be intelligently managed. The effectiveness of the system's thermal solution can be monitored using the *THERM* input. The ADT7460 also provides critical Thermal Protection to the system using the bidirectional *THERM* pin as an output to prevent system or component overheating.

FUNCTIONAL BLOCK DIAGRAM

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ADT7460—SPECIFICATIONS^{1,2,3,4} (T_A = T_{MIN} to T_{MAX}, V_{cc} = V_{MIN} to V_{MAX}, unless otherwise note

 $(T_A = T_{MIN}$ to T_{MAX} , $V_{CC} = V_{MIN}$ to V_{MAX} , unless otherwise noted.)

NOTES

 1 All voltages are measured with respect to GND, unless otherwise specified.

²Typicals are at T_A = 25°C and represent the most likely parametric norm.

³Logic inputs will accept input high voltages up to V_{MAX} even when the device is operating down to V_{MIN} .

⁴Timing specifications are tested at logic levels of V_{IL} = 0.8 V for a falling edge and V_{IH} = 2.0 V for a rising edge.

⁵The delay is the time between the round robin finishing one set of measurements and starting the next.

Specifications subject to change without notice.

Figure 1. Diagram for Serial Bus Timing

ABSOLUTE MAXIMUM RATINGS*

*Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

THERMAL CHARACTERISTICS 16-Lead QSOP Package: θ_{IA} = 150°C/W, θ_{IC} = 39°C/W

CAUTION

ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000 V readily accumulate on the human body and test equipment and can discharge without detection. Although the ADT7460 features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.

ORDERING GUIDE

PIN CONFIGURATION

PIN FUNCTION DESCRIPTIONS

FUNCTIONAL DESCRIPTION

General Description

The ADT7460 is a thermal monitor and multiple fan controller for any system requiring monitoring and cooling. The device communicates with the system via a serial System Management Bus. The serial bus controller has an optional address line for device selection (Pin 9), a serial data line for reading and writing addresses and data (Pin 16), and an input line for the serial clock (Pin 1). All control and programming functions of the ADT7460 are performed over the serial bus. In addition, two of the pins can be reconfigured as an *SMBALERT* output to indicate out-of-limit conditions.

Measurement Inputs

The device has three measurement inputs, one for voltage and two for temperature. It can also measure its own supply voltage and can measure ambient temperature with its on-chip temperature sensor.

Pin 14 is an analog input with an on-chip attenuator and is configured to monitor 2.5 V.

Power is supplied to the chip via Pin 3, and the system also monitors V_{CC} through this pin. In PCs, this pin is normally connected to a 3.3 V standby supply. This pin can, however, be connected to a 5 V supply and monitor it without overranging.

Remote temperature sensing is provided by the $D1\pm$ and $D2\pm$ inputs, to which diode-connected, external temperature-sensing transistors such as a 2N3904 or CPU thermal diode may be connected.

The ADC also accepts input from an on-chip band gap temperature sensor that monitors system ambient temperature.

Sequential Measurement

When the ADT7460 monitoring sequence is started, it cycles sequentially through the measurement of 2.5 V input and the temperature sensors. Measured values from these inputs are stored in Value registers. These can be read out over the serial bus, or can be compared with programmed limits stored in the Limit registers. The results of out-of-limit comparisons are stored in the Status registers, which can be read over the serial bus to flag out-of-limit conditions.

ADT7460 Address Selection

Pin 8 is the dual function PWM3/*ADDRESS ENABLE* pin. If Pin 8 is pulled low on power-up, the ADT7460 will read the state of Pin 9 (TACH4/ADDRESS SELECT/*THERM* pin) to determine the ADT7460's slave address. If Pin 8 is high on power-up, then the ADT7460 will default to SMBus slave address 0x2E. This function is described in more detail later.

INTERNAL REGISTERS OF THE ADT7460

A brief description of the ADT7460's principal internal registers is given below. More detailed information on the function of each register is given in Tables IV to XLI.

Configuration Registers

The Configuration registers provide control and configuration of the ADT7460, including alternate pinout functionality.

Address Pointer Register

This register contains the address that selects one of the other internal registers. When writing to the ADT7460, the first byte of data is always a register address, which is written to the Address Pointer Register.

Status Registers

These registers provide the status of each limit comparison and are used to signal out-of-limit conditions on the temperature, voltage, or fan speed channels. If Pin 14 is configured as *SMBALERT*, then this pin will assert low whenever an unmasked status bit gets set.

Interrupt Mask Registers

These registers allow each interrupt status event to be masked when Pin 14 is configured as an *SMBALERT* output.

Value and Limit Registers

The results of analog voltage input, temperature, and fan speed measurements are stored in these registers, along with their limit values.

Offset Registers

These registers allow each temperature channel reading to be offset by a twos complement value written to these registers.

T_{MIN} Registers

These registers program the starting temperature for each fan under Automatic Fan Speed Control.

TRANGE Registers

These registers program the temperature-to-fan speed control slope in Automatic Fan Speed Control mode for each PWM output.

Operating Point Registers

These registers define the target operating temperatures for each thermal zone when running under dynamic T_{MIN} control. This function allows the cooling solution to adjust dynamically in response to measured temperature and system performance.

Enhance Acoustics Registers

These registers allow each PWM output controlling fan to be tweaked to enhance the system's acoustics.

Typical Performance Characteristics–ADT7460

LEAKAGE RESISTANCE – M REMOTE TEMPERATURE ERROR – -**REMOTE TEMPERATURE ERROR - °C 15 10 –201 3.3 10 30 ¹⁰⁰ 0 –5 –10 –15 5** DXP TO GND **DXP TO V_{CC} (3.3V)**

 TPC 1. Temperature Error vs. Leakage Resistance

 TPC 2. Temperature Error vs. Capacitance between D+ and D–

 TPC 3. Remote Temperature Error vs. Actual Temperature

 TPC 4. Local Temperature Error vs. Actual Temperature

 TPC 5. Remote Temperature Error vs. Power Supply Noise Frequency

 TPC 7. Supply Current vs. Supply Voltage

 TPC 8. Remote Temperature Error vs. Differential Mode Noise Frequency

 TPC 6. Local Temperature Error vs. Power Supply Noise Frequency

 TPC 9. Remote Temperature Error vs. Common-Mode Noise Frequency

Figure 2. Recommended Implementation

RECOMMENDED IMPLEMENTATION

Configuring the ADT7460 as in Figure 2 allows the systems designer the following features:

- Two PWM outputs for fan control of up to three fans (the front and rear chassis fans are connected in parallel)
- Three TACH fan speed measurement inputs
- V_{CC} measured internally through Pin 3
- CPU temperature measured using Remote 1 temperature channel
- Ambient temperature measured through Remote 2 temperature channel
- Bidirectional *THERM* Pin. Allows Intel P4 *PROCHOT* Monitoring and can function as an overtemperature *THERM* output.
- *SMBALERT* system interrupt output

SERIAL BUS INTERFACE

Control of the ADT7460 is carried out using the serial System Management bus (SMBus). The ADT7460 is connected to this bus as a slave device, under the control of a master controller.

The ADT7460 has a 7-bit serial bus address. When the device is powered up with Pin 8 (PWM3/*Address Enable*) high, the ADT7460 will have a default SMBus address of 0101110 or 0x2E. If more than one ADT7460 is to be used in a system, then each ADT7460 should be placed in Address Select Mode by strapping Pin 8 low on power-up. The logic state of Pin 9 then determines the device's SMBus address.

The device address is sampled and latched on the first valid SMBus transaction, so any attempted addressing changes made thereafter will have no immediate effect.

Pin 8 State	Pin 9 State	Address		
Ω	Low (10 k Ω to GND)	0101100 (2Ch)		
Ω	High (10 k Ω pull-up)	0101101(2Dh)		
	Don't Care	0101110 (2Eh)		
		(default)		

Table I. Address Select Mode

Figure 3. Default SMBus Address = 0x2E

Figure 4. SMBus Address = $0x2C$ (Pin $9 = 0$)

The facility to make hardwired changes to the SMBus slave address allows the user to avoid conflicts with other devices sharing the same serial bus, for example, if more than one ADT7460 is used in a system.

The serial bus protocol operates as follows:

1. The master initiates data transfer by establishing a START condition, defined as a high to low transition on the serial data line SDA while the serial clock line SCL remains high. This indicates that an address/data stream will follow. All slave peripherals connected to the serial bus respond to the START condition and shift in the next eight bits, consisting of a 7-bit address (MSB first) plus a R/*W* bit, which determines the direction of the data transfer, i.e., whether data will be written to or read from the slave device.

Figure 5. SMBus Address = $0x2D$ (Pin $9 = 1$)

CARE SHOULD BE TAKEN TO ENSURE THAT PIN 8 (PWM3/ADDR_EN) IS EITHER TIED HIGH OR LOW. LEAVING PIN 8 FLOATING COULD CAUSE THE ADT7460 TO POWER UP WITH AN UNEXPECTED ADDRESS. NOTE THAT IF THE ADT7460 IS PLACED INTO ADDRESS SELECT MODE, PINS 8 AND 9 CAN BE USED AS THE ALTERNATE FUNC-TIONS (PWM3, TACH4/THERM) ONLY IF THE CORRECT CIRCUIT IS

MUXED IN AT THE CORRECT TIME Figure 6. Unpredictable SMBus Address if Pin 8

is Unconnected

REV. 0 –9– SUNSTAR自动化 http://www.sensor-ic.com/ TEL: 0755-83376489 FAX:0755-83376182 E-MAIL:szss20@163.comFigure 7. Writing a Register Address to the Address Pointer Register, Then Writing Data to the Selected Register

Figure 9. Reading Data from a Previously Selected Register

The peripheral whose address corresponds to the transmitted address responds by pulling the data line low during the low period before the ninth clock pulse, known as the Acknowledge Bit. All other devices on the bus now remain idle while the selected device waits for data to be read from or written to it. If the R/\overline{W} bit is a 0, then the master will write to the slave device. If the R/\overline{W} bit is a 1, the master will read from the slave device.

- 2. Data is sent over the serial bus in sequences of nine clock pulses, eight bits of data followed by an Acknowledge Bit from the slave device. Transitions on the data line must occur during the low period of the clock signal and remain stable during the high period, as a low to high transition when the clock is high may be interpreted as a STOP signal. The number of data bytes that can be transmitted over the serial bus in a single READ or WRITE operation is limited only by what the master and slave devices can handle.
- 3.When all data bytes have been read or written, stop conditions are established. In WRITE mode, the master will pull the data line high during the tenth clock pulse to assert a STOP condition. In READ mode, the master device will override the acknowledge bit by pulling the data line high during the low period before the ninth clock pulse. This is known as No Acknowledge. The master will then take the data line low during the low period before the tenth clock pulse, then high during the tenth clock pulse to assert a STOP condition.

Any number of bytes of data may be transferred over the serial bus in one operation, but it is not possible to mix read and write in one operation because the type of operation is determined at the beginning and cannot subsequently be changed without starting a new operation.

In the case of the ADT7460, write operations contain either one or two bytes, and read operations contain one byte and perform the following functions:

To write data to one of the device data registers or read data from it, the Address Pointer Register must be set so that the correct data register is addressed, then data can be written into that register or read from it. The first byte of a write operation always contains an address that is stored in the Address Pointer Register. If data is to be written to the device, then the write

operation contains a second data byte that is written to the register selected by the Address Pointer Register.

This is illustrated in Figure 7. The device address is sent over the bus followed by R/*W* being set to 0. This is followed by two data bytes. The first data byte is the address of the internal data register to be written to, which is stored in the Address Pointer Register. The second data byte is the data to be written to the internal data register.

When reading data from a register, there are two possibilities:

1. If the ADT7460's Address Pointer Register value is unknown or not the desired value, it is first necessary to set it to the correct value before data can be read from the desired data register. This is done by performing a write to the ADT7460 as before, but only the data byte containing the register address is sent as data is not to be written to the register. This is shown in Figure 8.

A read operation is then performed consisting of the serial bus address, R/*W* bit set to 1, followed by the data byte read from the data register. This is shown in Figure 9.

2. If the Address Pointer Register is known to be already at the desired address, data can be read from the corresponding data register without first writing to the Address Pointer Register, so Figure 8 can be omitted.

Notes

- 1. It is possible to *read* a data byte from a data register without first writing to the Address Pointer Register if the Address Pointer Register is already at the correct value. However, it is not possible to *write* data to a register without writing to the Address Pointer Register because the first data byte of a write is always written to the Address Pointer Register.
- 2. In Figures 7 to 9, the serial bus address is shown as the default value $01011(A1)(A0)$, where A1 and A0 are set by the Address Select Mode function previously defined.
- 3. In addition to supporting the Send Byte and Receive Byte protocols, the ADT7460 also supports the Read Byte protocol (see System Management Bus specifications Rev. 2.0 for more information).

4. If it is required to perform several read or write operations in succession, the master can send a repeat start condition instead of a stop condition to begin a new operation.

ADT7460 WRITE OPERATIONS

The SMBus specification defines several protocols for different types of read and write operations. The ones used in the ADT7460 are discussed below. The following abbreviations are used in the diagrams:

- S START
- P STOP
- R READ
- W WRITE
- A ACKNOWLEDGE
- *A* NO ACKNOWLEDGE

The ADT7460 uses the following SMBus write protocols:

Send Byte

In this operation, the master device sends a single command byte to a slave device as follows:

- 1. The master device asserts a START condition on SDA.
- 2. The master sends the 7-bit slave address followed by the write bit (low).
- 3. The addressed slave device asserts ACK on SDA.
- 4. The master sends a command code.
- 5. The slave asserts ACK on SDA.
- 6. The master asserts a STOP condition on SDA and the transaction ends.

For the ADT7460, the send byte protocol is used to write a register address to RAM for a subsequent single byte read from the same address. This is illustrated in Figure 10.

Figure 10. Setting a Register Address for Subsequent Read

If it is required to read data from the register immediately after setting up the address, the master can assert a repeat start condition immediately after the final ACK and carry out a single byte read without asserting an intermediate stop condition.

Write Byte

In this operation, the master device sends a command byte and one data byte to the slave device as follows:

- 1. The master device asserts a START condition on SDA.
- 2. The master sends the 7-bit slave address followed by the write bit (low).
- 3. The addressed slave device asserts ACK on SDA.
- 4. The master sends a command code.
- 5. The slave asserts ACK on SDA.
- 6. The master sends a data byte.
- 7. The slave asserts ACK on SDA.
- 8. The master asserts a STOP condition on SDA to end the transaction.

This is illustrated in Figure 11.

Figure 11. Single Byte Write to a Register

ADT7460 READ OPERATIONS

The ADT7460 uses the following SMBus read protocols:

Receive Byte

This is useful when repeatedly reading a single register. The register address needs to have been set up previously. In this operation, the master device receives a single byte from a slave device as follows:

- 1. The master device asserts a START condition on SDA.
- 2. The master sends the 7-bit slave address followed by the read bit (high).
- 3. The addressed slave device asserts ACK on SDA.
- 4. The master receives a data byte.
- 5. The master asserts NO ACK on SDA.
- 6. The master asserts a STOP condition on SDA and the transaction ends.

In the ADT7460, the receive byte protocol is used to read a single byte of data from a register whose address has previously been set by a send byte or write byte operation.

				Б,	
۱S	SLAVE W		REGISTER ADDRESS		

Figure 12. Single Byte Read from a Register

ALERT RESPONSE ADDRESS

Alert Response Address (ARA) is a feature of SMBus devices that allows an interrupting device to identify itself to the host when multiple devices exist on the same bus.

The *SMBALERT* output can be used as an interrupt output or can be used as an *SMBALERT*. One or more outputs can be connected to a common *SMBALERT* line connected to the master. If a device's *SMBALERT* line goes low, the following procedure occurs:

- 1. *SMBALERT* is pulled low.
- 2. Master initiates a read operation and sends the Alert Response Address ($ARA = 0001 100$). This is a general call address that must not be used as a specific device address.
- 3. The device whose *SMBALERT* output is low responds to the Alert Response Address, and the master reads its device address. The address of the device is now known, and it can be interrogated in the usual way.
- 4. If more than one device's *SMBALERT* output is low, the one with the lowest device address will have priority in accordance with normal SMBus arbitration.
- 5. Once the ADT7460 has responded to the Alert Response Address, the master must read the Status Registers and the *SMBALERT* will only be cleared if the error condition has gone away.

SMBUS TIMEOUT

The ADT7460 includes an SMBus Timeout feature. If there is no SMBus activity for 35 ms, the ADT7460 assumes that the bus is locked and releases the bus. This prevents the device from locking or holding the SMBus expecting data. Some SMBus controllers cannot handle the SMBus Timeout feature, so it can be disabled.

CONFIGURATION REGISTER 1 – Register 0x40 <6> TODIS = 0; SMBus Timeout ENABLED (default) **<6> TODIS = 1;** SMBus Timeout DISABLED

VOLTAGE MEASUREMENT INPUT

The ADT7460 has one external voltage measurement channel. It can also measure its own supply voltage, V_{CC} .

Pin 14 may be configured to measure a 2.5 V supply. The V_{CC} supply voltage measurement is carried out through the V_{CC} pin (Pin 3). Setting Bit 7 of Configuration Register 1 (Reg. 0x40) allows a 5 V supply to power the ADT7460 and be measured without overranging the V_{CC} measurement channel. The 2.5 V input can be used to monitor a chipset supply voltage in computer systems.

ANALOG-TO-DIGITAL CONVERTER

All analog inputs are multiplexed into the on-chip, successive approximation, analog-to-digital converter. This has a resolution of 10 bits. The basic input range is 0 V to 2.25 V but the input has built-in attenuators to allow measurement of 2.5 V without any external components. To allow for the tolerance of the supply voltage, the ADC produces an output of 3/4 full scale (decimal 768 or 300 hex) for the nominal input voltage and so has adequate headroom to deal with overvoltages.

INPUT CIRCUITRY

The internal structure for the 2.5 V analog input is shown in Figure 13. The input circuit consists of an input protection diode, an attenuator, plus a capacitor to form a first-order low-pass filter that gives the input immunity to high frequency noise.

VOLTAGE MEASUREMENT REGISTERS

Reg. $0x20$ **2.5 V Reading** = $0x00$ default

2.5 V LIMIT REGISTERS

Associated with the 2.5 V measurement channel is a high and low limit register. Exceeding the programmed high or low limit causes the appropriate Status bit to be set. Exceeding either limit can also generate **SMBALERT** interrupts.

Reg. $0x44$ 2.5 V Low Limit = $0x00$ default

Reg. $0x45$ 2.5 V High Limit = $0xFF$ default

Figure 13. Structure of Analog Inputs

Table II shows the input ranges of the analog inputs and output codes of the 10-bit ADC.

When the ADC is running, it samples and converts a voltage input in 711 µs and averages 16 conversions to reduce noise; a measurement takes nominally 11.38 ms.

Table II. 10-Bit A/D Output Code vs. V_{IN}

*The V_{CC} output codes listed assume that V_{CC} is 3.3 V. If V_{CC} input is reconfigured for 5 V operation (by setting Bit 7 of Configuration Register 1), then the V_{CC} output codes are the same as for the 5 $\mathrm{V_{IN}}$ column.

ADDITIONAL ADC FUNCTIONS FOR VOLTAGE MEASUREMENTS

A number of other functions are available on the ADT7460 to offer the systems designer increased flexibility:

Turn Off Averaging

For each voltage measurement read from a value register, 16 readings have actually been made internally and the results averaged before being placed into the value register. There may be an instance where you would like to speed up conversions. Setting Bit 4 of Configuration Register 2 (Reg. 0x73) turns averaging off. This effectively gives a reading 16 times faster $(711 \,\mu s)$, but the reading may be noisier.

Bypass Voltage Input Attenuator

Setting Bit 5 of Configuration Register 2 (Reg. 0x73) removes the attenuation circuitry from the 2.5 V input. This allows the user to directly connect external sensors or rescale the analog voltage measurement inputs for other applications. The input range of the ADC without the attenuators is 0 V to 2.25 V.

Single-Channel ADC Conversion

Setting Bit 6 of Configuration Register 2 (Reg. 0x73) places the ADT7460 into Single-Channel ADC Conversion Mode. In this mode, the ADT7460 can be made to read a single voltage channel only. If the internal ADT7460 clock is used, the selected input will be read every 711 µs. The appropriate ADC channel is selected by writing to Bits <7:5> of the TACH1 Minimum High Byte Register (0x55).

Configuration Register 2 (Reg. 0x73)

<4> = 1 Averaging Off

- **<5> = 1** Bypass Input Attenuators
- **<6> = 1** Single-Channel Convert Mode

TACH1 Minimum High Byte (Reg. 0x55)

<7:5> Selects ADC Channel for Single-Channel Convert Mode

TEMPERATURE MEASUREMENT SYSTEM Local Temperature Measurement

The ADT7460 contains an on-chip band gap temperature sensor whose output is digitized by the on-chip 10-bit ADC. The 8-bit MSB temperature data is stored in the Local Temp Register (Address 26h). As both positive and negative temperatures can be measured, the temperature data is stored in twos complement format, as shown in Table III. Theoretically, the temperature sensor and ADC can measure temperatures from -128° C to $+127^{\circ}$ C with a resolution of 0.25° C. However, this exceeds the operating temperature range of the device, so local temperature measurements outside this range are not possible.

Remote Temperature Measurement

The ADT7460 can measure the temperature of two remote diode sensors or diode-connected transistors connected to Pins 12 and 13, or 10 and 11.

The forward voltage of a diode or diode-connected transistor operated at a constant current exhibits a negative temperature coefficient of about -2 mV/ \degree C. Unfortunately, the absolute value of V_{BE} varies from device to device and individual calibration is required to null this out, so the technique is unsuitable for mass production. The technique used in the ADT7460 is to measure the change in V_{BE} when the device is operated at two different currents.

This is given by:

$$
\Delta V_{BE} = KT/q \times \ln(N)
$$

where:

K is Boltzmann's constant

q is charge on the carrier

T is absolute temperature in Kelvins

N is ratio of the two currents.

Figure 14 shows the input signal conditioning used to measure the output of a remote temperature sensor. This figure shows the external sensor as a substrate transistor, provided for temperature monitoring on some microprocessors. It could equally well be a discrete transistor such as a 2N3904.

Figure 14. Signal Conditioning for Remote Diode Temperature Sensors

If a discrete transistor is used, the collector will not be grounded, and should be linked to the base. If a PNP transistor is used, the base is connected to the D– input and the emitter to the D+ input. If an NPN transistor is used, the emitter is connected to the D– input and the base to the D+ input. Figure 15 shows how to connect the ADT7460 to an NPN or PNP transistor for temperature measurement. To prevent ground noise from interfering with the measurement, the more negative terminal of the sensor is not referenced to ground but is biased above ground by an internal diode at the D– input.

To measure ΔV_{BE} , the sensor is switched between operating currents of I and $N \times I$. The resulting waveform is passed through a 65 kHz low-pass filter to remove noise, and to a chopperstabilized amplifier that performs the functions of amplification and rectification of the waveform to produce a dc voltage proportional to ΔV_{BE} . This voltage is measured by the ADC to give a temperature output in 10-bit, twos complement format. To further reduce the effects of noise, digital filtering is performed by averaging the results of 16 measurement cycles. A remote temperature measurement takes nominally 25.5 ms. The results of remote temperature measurements are stored in 10-bit, twos complement format, as illustrated in Table III. The extra resolution for the temperature measurements is held in the Extended Resolution Register 2 (Reg. 0x77). This gives temperature readings with a resolution of 0.25° C.

Table III. Temperature Data Format

Temperature	Digital Output (10-Bit)*			
-128 °C	1000 0000 00			
-125° C	1000 0011 00			
-100° C	1001 1100 00			
-75° C	1011 0101 00			
-50° C	1100 1110 00			
-25° C	1110 0111 00			
-10° C	1111 0110 00			
0° C	0000 0000 00			
$+10.25^{\circ}$ C	0000 1010 01			
$+25.5^{\circ}$ C	0001 1001 10			
$+50.75^{\circ}$ C	0011 0010 11			
$+75^{\circ}$ C	0100 1011 00			
$+100^{\circ}$ C	0110 0100 00			
$+125^{\circ}$ C	0111 1101 00			
$+127^{\circ}$ C	0111 1111 00			

*Bold denotes 2 LSBs of measurement in Extended Resolution Register 2 (Reg. $0x77$) with 0.25° C resolution.

Figure 15a. Measuring Temperature Using an NPN Transistor

Figure 15b. Measuring Temperature Using a PNP Transistor

Nulling Out Temperature Errors

As CPUs run faster, it is getting more difficult to avoid high frequency clocks when routing the D+, D– traces around a system board. Even when recommended layout guidelines are followed, there may still be temperature errors attributed to noise being coupled onto the $D+/D-$ lines. High frequency noise generally has the effect of giving temperature measurements that are too high by a constant amount. The ADT7460 has temperature offset registers at addresses 0x70, 0x72 for the Remote 1 and Remote 2 temperature channels. By doing a one-time calibration of the system, one can determine the offset caused by system board noise and null it out using the offset registers. The offset registers automatically add a twos complement 8-bit reading to every temperature measurement. The LSB adds 0.25°C offset to the temperature reading so the 8-bit register effectively allows temperature offsets of up to $\pm 32^{\circ}$ C with a resolution of 0.25 $^{\circ}$ C. This ensures that the readings in the temperature measurement registers are as accurate as possible.

Temperature Offset Registers

Reg. 0x70 Remote 1 Temp Offset = $0x00$ (0 $^{\circ}$ C default) Reg. 0x71 Local Temp Offset = $0x00$ (0 $^{\circ}$ C default) Reg. 0x72 Remote 2 Temp Offset = $0x00$ (0 \degree C default)

Temperature Measurement Registers

Reg. 0x25 **Remote 1 Temperature** = 0x80 default Reg. 0x26 **Local Temperature** = 0x80 default Reg. 0x27 **Remote 2 Temperature** = 0x80 default

Reg. 0x77 **Extended Resolution 2** = 0x00 default **<7:6> TDM2** = Remote 2 Temperature LSBs **<5:4> LTMP** = Local Temperature LSBs **<3:2> TDM1** = Remote 1 Temperature LSBs

Temperature Measurement Limit Registers

Associated with each temperature measurement channel are high and low limit registers. Exceeding the programmed high or low limit causes the appropriate Status bit to be set. Exceeding either limit can also generate *SMBALERT* interrupts.

Reg. 0x4E **Remote 1 Temp Low Limit** = 0x81 default

Reg. 0x4F **Remote 1 Temp High Limit** = 0x7F default

Reg. 0x50 **Local Temp Low Limit** = 0x81 default

Reg. 0x51 **Local Temp High Limit** = 0x7F default

Reg. 0x52 **Remote 2 Temp Low Limit** = 0x81 default Reg. 0x53 **Remote 2 Temp High Limit** = 0x7F default

Reading Temperature from the ADT7460

It is important to note that temperature can be read from the ADT7460 as an 8-bit value (with 1°C resolution), or as a 10-bit value (with 0.25°C resolution). If only 1°C resolution is required, the temperature readings can be read back at any time and in no particular order.

If the 10-bit measurement is required, this involves a 2-register read for each measurement. The Extended Resolution Register (Reg. 0x77) should be read first. This causes all temperature reading registers to be frozen until all temperature reading registers have been read from. This prevents an MSB reading from being updated while its two LSBs are being read, and vice versa.

ADDITIONAL ADC FUNCTIONS FOR TEMPERATURE MEASUREMENT

A number of other functions are available on the ADT7460 to offer the systems designer increased flexibility:

Turn Off Averaging

For each temperature measurement read from a value register, 16 readings have actually been made internally and the results averaged before being placed into the value register. Sometimes it may be necessary to take a very fast measurement, e.g., of CPU temperature. Setting Bit 4 of Configuration Register 2 (Reg. 0x73) turns averaging off. This takes a reading every 15.5 ms. Each remote temperature measurement takes 4 ms and the local temperature measurement takes 1.4 ms.

Single-Channel ADC Conversions

Setting Bit 6 of Configuration Register 2 (Reg. 0x73) places the ADT7460 into single-channel ADC Conversion Mode. In this mode, the ADT7460 can be made to read a single temperature channel only. The appropriate ADC channel is selected by writing to Bits <7:5> of TACH1 Minimum High Byte Register (0x55).

Configuration Register 2 (Reg. 0x73)

<4> = 1 Averaging Off

<6> = 1 Single-Channel Convert Mode

TACH1 Minimum High Byte (Reg. 0x55)

<7:5> Selects ADC Channel for Single-Channel Convert Mode

Overtemperature Events

Overtemperature events on any of the temperature channels can be detected and dealt with automatically in Automatic Fan Speed Control Mode. Registers 0x6A–0x6C are the *THERM* limits. When a temperature exceeds its *THERM* limit, all fans will run at 100% duty cycle. The fans will stay running at 100% until the temperature drops below *THERM* – Hysteresis (this can be disabled by setting the Boost bit in Configuration Register 3, Bit 2, Register 0x78). The hysteresis value for that *THERM* limit is the value programmed into Registers 0x6D, $0x6E$ (Hysteresis registers). The default hysteresis value is $4^{\circ}C$.

Figure 16. THERM Limit Operation

LIMITS, STATUS REGISTERS, AND INTERRUPTS Limit Values

Associated with each measurement channel on the ADT7460 are high and low limits. These can form the basis of system status monitoring: a Status bit can be set for any out-of-limit condition and detected by polling the device. Alternatively, *SMBALERT* interrupts can be generated to flag a processor or microcontroller of out-of-limit conditions.

8-Bit Limits

The following is a list of 8-bit limits on the ADT7460:

Voltage Limit Registers

Reg. $0x44$ 2.5 V Low Limit = $0x00$ default

Reg. 0x45 **2.5 V High Limit** = 0xFF default

Reg. $0x48$ V_{CC} **Low Limit** = $0x00$ default

Reg. 0x49 V_{CC} High Limit = 0xFF default

Temperature Limit Registers

Reg. 0x4E **Remote 1 Temp Low Limit** = 0x81 default

Reg. 0x4F **Remote 1 Temp High Limit** = 0x7F default

Reg. 0x6A **Remote 1** *THERM* **Limit** = 0x64 default

Reg. 0x50 **Local Temp Low Limit** = 0x81 default

Reg. 0x51 **Local Temp High Limit** = 0x7F default

Reg. 0x6B **Local** *THERM* **Limit** = 0x64 default

Reg. 0x52 **Remote 2 Temp Low Limit** = 0x81 default

Reg. 0x53 **Remote 2 Temp High Limit** = 0x7F default

Reg. 0x6C **Remote 2** *THERM* **Limit** = 0x64 default

Therm Limit Register

Reg. 0x7A *THERM* **Timer Limit** = 0x00 default

16-Bit Limits

The Fan TACH measurements are 16-bit results. The Fan TACH limits are also 16 bits, consisting of a High Byte and Low Byte. Since fans running under speed or stalled are normally the only conditions of interest, only High Limits exist for Fan TACHs. Since Fan TACH period is actually being measured, exceeding the limit indicates a slow or stalled fan.

Fan Limit Registers

Reg. 0x54 **TACH1 Minimum Low Byte** = 0xFF default Reg. 0x55 **TACH1 Minimum High Byte** = 0xFF default Reg. 0x56 **TACH2 Minimum Low Byte** = 0xFF default Reg. 0x57 **TACH2 Minimum High Byte** = 0xFF default Reg. 0x58 **TACH3 Minimum Low Byte** = 0xFF default Reg. 0x59 **TACH3 Minimum High Byte** = 0xFF default Reg. 0x5A **TACH4 Minimum Low Byte** = 0xFF default Reg. 0x5B **TACH4 Minimum High Byte** = 0xFF default

Out-of-Limit Comparisons

Once all limits have been programmed, the ADT7460 can be enabled for monitoring. The ADT7460 will measure all parameters in round-robin format and set the appropriate Status bit for out-of-limit conditions. Comparisons are done differently depending on whether the measured value is being compared to a high or low limit.

HIGH LIMIT: > COMPARISON PERFORMED LOW LIMIT: < OR = COMPARISON PERFORMED

Figure 17. Temperature > Low Limit: No \overline{INT}

Figure 18. Temperature = Low Limit: INT Occurs

Figure 20. Temperature > High Limit: \overline{INT} Occurs

Analog Monitoring Cycle Time

The analog monitoring cycle begins when a 1 is written to the Start bit (Bit 0) of Configuration Register 1 (Reg 0x40). The ADC measures each analog input in turn and as each measurement is completed, the result is automatically stored in the appropriate value register. This round-robin monitoring cycle continues unless disabled by writing a 0 to Bit 0 of Configuration Register 1.

As the ADC will normally be left to free-run in this manner, the time taken to monitor all the analog inputs will normally not be of interest, as the most recently measured value of any input can be read out at any time.

For applications where the monitoring cycle time is important, it can easily be calculated.

The total number of channels measured is:

Two supply voltage inputs $(2.5 V and V_{CC})$

Local temperature

Two remote temperatures

As mentioned previously, the ADC performs round-robin conversions and takes 11.38 ms for each voltage measurement, 12 ms for a local temperature reading, and 25.5 ms for each remote temperature reading.

The total monitoring cycle time for averaged voltage and temperature monitoring is therefore nominally:

 $(2 \times 11.38) + 12 + (2 \times 25.5) = 85.76$ ms

The round robin starts again 35 ms later. Therefore all channels are measured approximately every 120 ms*.*

Fan TACH measurements are made in parallel and are not synchronized with the analog measurements in any way.

Status Registers

The results of limit comparisons are stored in Status Registers 1 and 2. The Status Register bit for each channel reflects the status of the last measurement and limit comparison on that channel. If a measurement is within limits, the corresponding status register bit will be cleared to 0. If the measurement is out-of-limits, the corresponding status register bit will be set to 1.

The state of the various measurement channels may be polled by reading the Status Registers over the serial bus. In Bit 7 (OOL) of Status Register 1 (Reg. 0x41), 1 means that an out-of-limit event has been flagged in Status Register 2. This means that you need only read Status Register 2 when this bit is set. Alternatively, Pin 5 or Pin 14 can be configured as an *SMBALERT* output. This will automatically notify the system supervisor of an out-of-limit condition. Reading the Status registers clears the appropriate status bit as long as the error condition that caused the interrupt has cleared. Status Register bits are "sticky." Whenever a Status bit gets set, indicating an out-of-limit condition, it will remain set even if the event that caused it has gone away (until read). The only way to clear the status bit is to read the Status Register after the event has gone away. Interrupt Status Mask Registers (Reg. 0x74, 0x75) allow individual interrupt sources to be masked from causing an *SMBALERT*. However, if one of these masked interrupt sources goes out-of-limit, its associated status bit will get set in the Interrupt Status Registers.

Figure 21. Status Register 1

Status Register 1 (Reg. 0x41)

Bit 7 (OOL) = 1, denotes a bit in Status Register 2 is set and Status Register 2 should be read.

Bit 6 (R2T) = 1, Remote 2 Temp High or Low Limit has been exceeded.

Bit 5 (LT) = 1, Local Temp High or Low Limit has been exceeded.

Bit 4 (R1T) = 1, Remote 1 Temp High or Low Limit has been exceeded.

Bit 3 = Unused

Bit 2 (V_{CC}) = 1, V_{CC} High or Low Limit has been exceeded.

Bit 1 = Unused

Bit 0 (2.5 V) = 1, 2.5 V High or Low Limit has been exceeded.

Figure 22. Status Register 2

Status Register 2 (Reg. 0x42)

Bit 7 (D2) = 1, indicates an open or short on $D2+/D2$ – inputs.

Bit 6 (D1) = 1, indicates an open or short on $D2+/D2$ – inputs.

Bit 5 (F4P) = 1, indicates Fan 4 has dropped below minimum speed. Alternatively, indicates that *THERM* Timer limit has been exceeded if the *THERM* Timer function is used.

Bit 4 (FAN3) = 1, indicates Fan 3 has dropped below minimum speed.

Bit 3 (FAN2) = 1, indicates Fan 2 has dropped below minimum speed.

Bit 2 (FAN1) = 1, indicates Fan 1 has dropped below minimum speed.

Bit 1 (OVT) = 1, indicates that a *THERM* overtemperature limit has been exceeded.

Bit 0 = Unused

SMBALERT Interrupt Behavior

The ADT7460 can be polled for status, or an *SMBALERT* interrupt can be generated for out-of-limit conditions. It is important to note how the *SMBALERT* output and status bits behave when writing Interrupt Handler software.

Figure 23. SMBALERT and Status Bit Behavior

Figure 23 shows how the *SMBALERT* output and "sticky" status bits behave. Once a limit is exceeded, the corresponding status bit gets set to 1. The status bit remains set until the error condition subsides and the Status Register gets read. The status bits are referred to as "sticky" since they remain set until read by software. This ensures that an out-of-limit event cannot be missed if software is polling the device periodically. Note that the *SMBALERT* output remains low for the entire duration that a reading is out-of-limit and until the Status Register has been read. This has implications on how software handles the interrupt.

HANDLING SMBALERT INTERRUPTS

To prevent the system from being tied up servicing interrupts, it is recommend to handle the *SMBALERT* interrupt as follows:

- 1. Detect the *SMBALERT* assertion.
- 2. Enter the interrupt handler.
- 3. Read the Status Registers to identify the interrupt source.
- 4. Mask the interrupt source by setting the appropriate Mask bit in the Interrupt Mask Registers (Reg. 0x74, 0x75).
- 5. Take the appropriate action for a given interrupt source.
- 6. Exit the Interrupt Handler.
- 7. Periodically poll the Status Registers. If the interrupt status bit has cleared, reset the corresponding Interrupt Mask Bit to 0. This will cause the *SMBALERT* output and status bits to behave as shown in Figure 24.

Figure 24. How Masking the Interrupt Source Affects SMBALERT Output

Masking Interrupt Sources

Interrupt Mask Registers 1 and 2 are located at Addresses 0x74 and 0x75. These allow individual interrupt sources to be masked out to prevent *SMBALERT* interrupts. Note that masking an interrupt source only prevents the *SMBALERT* output from being asserted; the appropriate Status bit will get set as normal.

Interrupt Mask Register 1 (Reg. 0x74)

Bit 7 (OOL) = 1, masks *SMBALERT* for any alert condition flagged in Status Register 2.

Bit 6 (R2T) = 1, masks *SMBALERT* for Remote 2 Temperature.

Bit 5 (LT) = 1, masks *SMBALERT* for Local Temperature.

Bit 4 (R1T) = 1, masks *SMBALERT* for Remote 1 Temperature.

Bit 3 = Unused

Bit 2 (V_{CC}) = 1, masks $\overline{\text{SMBALERT}}$ **for V_{CC} channel.**

Bit 1 = Unused

Bit 0 (2.5 V) = 1, masks *SMBALERT* for 2.5 V channel.

Interrupt Mask Register 2 (Reg. 0x75)

Bit 7 (D2) = 1, masks *SMBALERT* for Diode 2 errors.

Bit 6 (D1) = 1, masks *SMBALERT* for Diode 1 errors.

Bit 5 (FAN4) = 1, masks *SMBALERT* for Fan 4 failure. If the TACH4 pin is being used as the *THERM* input, this bit masks *SMBALERT* for a *THERM* event.

Bit 4 (FAN3) = 1, masks *SMBALERT* for Fan 3.

Bit 3 (FAN2) = 1, masks *SMBALERT* for Fan 2.

Bit 2 (FAN1) = 1, masks *SMBALERT* for Fan 1.

Bit 1 (OVT) = 1, masks *SMBALERT* for overtemperature (exceeding *THERM* limits).

Bit 0 = Unused

Enabling the SMBALERT Interrupt Output

The *SMBALERT* interrupt function is disabled by default. Pin 5 or Pin 14 can be reconfigured as an *SMBALERT* output to signal out-of-limit conditions.

CONFIGURING PIN 5 AS SMBALERT OUTPUT

CONFIGURING PIN 14 AS SMBALERT OUTPUT REGISTER BIT SETTING

Therm Input

The ADT7460 has an internal timer to measure *THERM* assertion time. For example, the *THERM* input may be connected to the *PROCHOT* output of a Pentium 4 CPU and measure system performance. The *THERM* input may also be connected to the output of a trip point temperature sensor.

The timer is started on the assertion of the ADT7460's *THERM* input, and stopped on the negation of the pin. The timer counts *THERM* times cumulatively, i.e., the timer resumes counting on the next *THERM* assertion. The *THERM* timer will continue to accumulate *THERM* assertion times until the timer is read (it is cleared on read) or until it reaches full scale. If the counter reaches full scale, it will stop at that reading until cleared.

The 8-bit *THERM* Timer register (Reg. 0x79) is designed such that Bit 0 will get set to 1 on the first *THERM* assertion. Once the cumulative *THERM* assertion time has exceeded 45.52 ms, Bit 1 of the *THERM* timer gets set, and Bit 0 now becomes the LSB of the timer with a resolution of 22.76 ms.

Figure 25. Understanding the THERM Timer

Figure 25 illustrates how the *THERM* timer behaves as the *THERM* input is asserted and negated. Bit 0 gets set on the first *THERM* assertion detected. This bit remains set until such time as the cumulative *THERM* assertions exceed 45.52 ms. At this time, Bit 1 of the *THERM* timer gets set, and Bit 0 is cleared. Bit 0 now reflects timer readings with a resolution of 22.76 ms.

When using the *THERM* timer, be aware of the following:

After a *THERM* timer read (Reg. 0x79):

- a) The contents of the timer get cleared on read.
- b) The F4P bit (Bit 5) of Status Register 2 needs to be cleared (assuming the *THERM* limit has been exceeded).

If the *THERM* timer is read during a *THERM* assertion, then the following will happen:

a) The contents of the timer are cleared.

- b) Bit 0 of the *THERM* timer is set to 1 (since a *THERM* assertion is occurring).
- c) The *THERM* timer increments from zero.
- d) If the $\overline{\text{THERM}}$ limit (Reg. 0x7A) = 0x00, then the F4P bit gets set.

Generating SMBALERT Interrupts from THERM Events The ADT7460 can generate *SMBALERT*s when a programmable *THERM* limit has been exceeded. This allows the systems designer to ignore brief, infrequent *THERM* assertions, while capturing longer *THERM* events. Register 0x7A is the

THERM Limit Register. This 8-bit register allows a limit from 0 seconds (first *THERM* assertion) to 5.825 seconds to be set before an *SMBALERT* is generated. The *THERM* Timer value is compared with the contents of the *THERM* Limit Register. If the *THERM* Timer value exceeds the *THERM* Limit value, then the F4P bit (Bit 5) of Status Register 2 gets set, and an *SMBALERT* is generated. Note that the F4P bit (Bit 5) of Mask Register 2 (Reg. 0x75) will mask out *SMBALERT*s if this bit is set to 1, although the F4P bit of Interrupt Status Register 2 will still get set if the *THERM* Limit is exceeded.

Figure 26 is a Functional Block Diagram of the *THERM* timer, limit, and associated circuitry. Writing a value of 0x00 to the *THERM* Limit Register (Reg. 0x7A) causes *SMBALERT* to be generated on the first *THERM* assertion. A *THERM* Limit value of 0x01 generates an *SMBALERT* once cumulative *THERM* assertions exceed 45.52 ms.

Figure 26. Functional Diagram of ADT7460's THERM Monitoring Circuitry

Configuring the Desired THERM Behavior

1. Configure the THERM input.

Setting Bit 1 (PHOT) of Configuration Register 3 (Reg. 0x78) enables the *THERM* monitoring function.

2. Select the desired fan behavior for THERM events. Setting Bit 2 (BOOST bit) of Configuration Register 3 (Reg. 0x78) causes all fans to run at 100% duty cycle whenever *THERM* gets asserted. This allows fail-safe system cooling. If this bit $= 0$, the fans will run at their current settings and will not be affected by *THERM* events.

3. Select whether THERM events should generate SMBALERT interrupts.

Bit 5 (F4P) of Mask Register 2 (Reg. 0x75), when set, masks out *SMBALERT*s when the *THERM* limit value gets exceeded. This bit should be cleared if *SMBALERT*s based on *THERM* events are required.

4. Select a suitable THERM limit value.

This value determines whether an *SMBALERT* is generated on the first *THERM* assertion, or only if a cumulative *THERM* assertion time limit is exceeded. A value of 0x00 causes an *SMBALERT* to be generated on the first *THERM* assertion.

5. Select a THERM monitoring time.

This is how often OS or BIOS level software checks the *THERM* timer. For example, BIOS could read the *THERM* timer once an hour to determine the cumulative *THERM* assertion time. If, for example, the total *THERM* assertion time is <22.76 ms in Hour 1, >182.08 ms in Hour 2, and >5.825 s in Hour 3, this can indicate that system performance is degrading significantly since *THERM* is asserting more frequently on an hourly basis.

Alternatively, OS or BIOS level software can time-stamp when the system is powered on. If an *SMBALERT* is generated due to the *THERM* limit being exceeded, another time-stamp can be taken. The difference in time can be calculated for a fixed

THERM limit time. For example, if it takes one week for a *THERM* limit of 2.914 s to be exceeded and the next time it only takes 1 hour, then this is an indication of a serious degradation in system performance.

Configuring the ADT7460 THERM Pin as an Output

In addition to the ADT7460 being able to monitor *THERM* as an input, the ADT7460 can optionally drive *THERM* low as an output. The user can preprogram system critical thermal limits. If the temperature exceeds a thermal limit by 0.25°C, *THERM* will assert low. If the temperature is still above the thermal limit on the next monitoring cycle, *THERM* will stay low. *THERM* will remain asserted low until the temperature is equal to or below the thermal limit. Since the temperature for that channel is only measured every monitoring cycle, once *THERM* asserts it is guaranteed to remain low for at least one monitoring cycle.

The *THERM* pin can be configured to assert low if the Remote 1, Local, or Remote 2 Temperature *THERM* Limits get exceeded by 0.25°C. The *THERM* Limit Registers are at locations 0x6A, 0x6B, and 0x6C respectively. Setting Bit 3 of Registers 0x5F, 0x60, and 0x61 enables the *THERM* output feature for the Remote 1, Local, and Remote 2 Temperature channels, respectively. Figure 27 shows how the *THERM* pin asserts low as an output in the event of a critical overtemperature.

Figure 27. Asserting THERM as an Output, Based on Tripping THERM Limits

FAN DRIVE USING PWM CONTROL

The ADT7460 uses Pulsewidth Modulation (PWM) to control fan speed. This relies on varying the duty cycle (or on/off ratio) of a square wave applied to the fan to vary the fan speed. The external circuitry required to drive a fan using PWM control is extremely simple. A single NMOSFET is the only drive device required. The specifications of the MOSFET depend on the maximum current required by the fan being driven. Typical notebook fans draw a nominal 170 mA, and so SOT devices can be used where board space is a concern. In desktops, fans can typically draw 250 mA–300 mA each. If you drive several fans in parallel from a single PWM output or drive larger server fans, the MOSFET will need to handle the higher current requirements. The only other stipulation is that the MOSFET should have a gate voltage drive, V_{GS} < 3.3 V for direct interfacing to the PWM_OUT pin. V_{GS} can be greater than 3.3 V as long as the pull-up on the gate is tied to 5 V. The MOSFET should also have a low on resistance to ensure that there is not significant voltage drop across the FET. This would reduce the voltage applied across the fan and therefore the maximum operating speed of the fan.

Figure 28 shows how a 3-wire fan may be driven using PWM control.

Figure 28. Driving a 3-Wire Fan Using an N-Channel MOSFET

Figure 28 uses a 10 kΩ pull-up resistor for the TACH signal. This assumes that the TACH signal is open-collector from the fan. In all cases, the TACH signal from the fan *must* be kept below 5 V maximum to prevent damaging the ADT7460. If in doubt as to whether the fan used has an open-collector or totem pole TACH output, use one of the input signal conditioning circuits shown in the Fan Speed Measurement section of the data sheet.

Figure 29 shows a fan drive circuit using an NPN transistor such as a general-purpose MMBT2222. While these devices are inexpensive, they tend to have much lower current handling capabilities and higher on-resistance than MOSFETs. When choosing a transistor, care should be taken to ensure that it meets the fan's current requirements.

Ensure that the base resistor is chosen such that the transistor is saturated when the fan is powered on.

Figure 29. Driving a 3-Wire Fan Using an NPN Transistor

Driving Two Fans from PWM3

Note that the ADT7460 has four TACH inputs available for fan speed measurement, but only three PWM drive outputs. If a fourth fan is being used in the system, it should be driven from the PWM3 output in parallel with the third fan. Figure 30 shows how to drive two fans in parallel using low cost NPN transistors. Figure 31 is the equivalent circuit using the NDT3055L MOSFET. Note that since the MOSFET can handle up to 3.5 A, it is simply a matter of connecting another fan directly in parallel with the first.

Care should be taken in designing drive circuits with transistors and FETs to ensure that the PWM Pins are not required to source current, and that they sink less than the 8 mA maximum current specified on the data sheet.

Driving up to Three Fans From PWM2

TACH measurements for fans are synchronized to particular PWM channels, e.g., TACH1 is synchronized to PWM1. TACH3 and TACH4 are both synchronized to PWM3, so PWM3 can drive two fans. Alternatively, PWM2 can be programmed to synchronize TACH2, TACH3, and TACH4 to the PWM2 output. This allows PWM2 to drive two or three fans. In this case, the drive circuitry looks the same as shown in Figures 30 and 31. The SYNC bit in Register 0x62 enables this function.

<4> (SYNC) ENHANCE ACOUSTICS REG 1 (0x62) SYNC = 1 Synchronizes TACH2, TACH3, and TACH4 to PWM2.

Figure 30. Interfacing Two Fans in Parallel to the PWM3 Output Using Low Cost NPN Transistors

Figure 31. Interfacing Two Fans in Parallel to the PWM3 Output Using a Single N-Channel MOSFET

Driving 2-Wire Fans

Figure 32 shows how a 2-wire fan may be connected to the ADT7460. This circuit allows the speed of a 2-wire fan to be measured, even though the fan has no dedicated TACH signal. A series resistor, RSENSE, in the fan circuit converts the fan commutation pulses into a voltage. This is ac-coupled into the ADT7460 through the 0.01 µF capacitor. On-chip signal conditioning allows accurate monitoring of fan speed. The value of R_{SENSE} chosen depends upon the programmed input threshold and the current drawn by the fan. For fans drawing approximately 200 mA, a 2 Ω R_{SENSE} value is suitable when the threshold is programmed as 40 mV. For fans that draw more current, such as larger desktop or server fans, R_{SENSE} may be reduced for the same programmed threshold. The smaller the threshold programmed the better, since more voltage will be developed across the fan and the fan will spin faster. Figure 33 shows a typical plot of the sensing waveform at a TACH/AIN pin. The most important thing is that the voltage spikes (either negative going or positive going) are more than 40 mV in amplitude. This allows fan speed to be reliably determined.

Figure 32. Driving a 2-Wire Fan

Figure 33. Fan Speed Sensing Waveform at TACH/AIN Pin

LAYING OUT 2-WIRE AND 3-WIRE FANS

Figure 34 shows how to lay out a common circuit arrangement for 2-wire and 3-wire fans. Some components will not be populated, depending on whether a 2-wire or 3-wire fan is being used.

TACH Inputs

Pins 4, 6, 7, and 9 are open-drain TACH inputs intended for fan speed measurement.

Signal conditioning in the ADT7460 accommodates the slow rise and fall times typical of fan tachometer outputs. The maximum input signal range is 0 V to 5 V, even where V_{CC} is less than 5 V. In the event that these inputs are supplied from fan outputs that exceed 0 V to 5 V, either resistive attenuation of the fan signal or diode clamping must be included to keep inputs within an acceptable range.

Figures 35a to 35d show circuits for most common fan TACH outputs.

If the fan TACH output has a resistive pull-up to V_{CC} , it can be connected directly to the fan input, as shown in Figure 35a.

Figure 35a. Fan with TACH Pull-Up to $+V_{CC}$

If the fan output has a resistive pull-up to 12 V (or other voltage greater than 5 V), the fan output can be clamped with a Zener diode, as shown in Figure 35b. The Zener diode voltage should be chosen so that it is greater than V_{IH} of the TACH input but less than 5 V, allowing for the voltage tolerance of the Zener. A value of between 3 V and 5 V is suitable.

***CHOOSE ZD1 VOLTAGE APPROX 0.8 VCC**

Figure 35b. Fan with TACH Pull-Up to Voltage > 5 V, e.g., 12 V, Clamped with Zener Diode

If the fan has a strong pull-up (less than 1 k Ω) to 12 V or a totem-pole output, then a series resistor can be added to limit the Zener current, as shown in Figure 35c. Alternatively, a resistive attenuator may be used, as shown in Figure 35d.

R1 and R2 should be chosen such that:

$$
2\,\text{V}<\!V_{PULLUP}\times R2\,/\,(R_{PULLUP}+R1+R2)\!<5\,\text{V}
$$

The fan inputs have an input resistance of nominally 160 kΩ to ground, so this should be taken into account when calculating resistor values.

With a pull-up voltage of 12 V and pull-up resistor less than 1 k Ω , suitable values for R1 and R2 would be 100 kΩ and 47 kΩ. This will give a high input voltage of 3.83 V.

Figure 35c. Fan with Strong TACH Pull-Up to $> V_{CC}$ or Totem-Pole Output, Clamped with Zener and Resistor

Fan Speed Measurement

The fan counter does not count the fan TACH output pulses directly because the fan speed may be less than 1000 RPM and it would take several seconds to accumulate a reasonably large and accurate count. Instead, the period of the fan revolution is measured by gating an on-chip 90 kHz oscillator into the input of a 16-bit counter for N periods of the fan TACH output (Figure 36), so the accumulated count is actually proportional to the fan tachometer period and inversely proportional to the fan speed.

ADT7460

Figure 36. Fan Speed Measurement

N, the number of pulses counted, is determined by the settings of Register 0x7B (Fan Pulses Per Revolution Register). This register contains two bits for each fan, allowing one, two (default), three, or four TACH pulses to be counted.

Fan Speed Measurement Registers

The Fan Tachometer Readings are 16-bit values consisting of a 2-byte read from the ADT7460.

Reg. 0x28 **TACH1 Low Byte** = 0x00 default Reg. 0x29 **TACH1 High Byte** = 0x00 default Reg. 0x2A **TACH2 Low Byte** = 0x00 default Reg. 0x2B **TACH2 High Byte** = 0x00 default Reg. 0x2C **TACH3 Low Byte** = 0x00 default Reg. 0x2D **TACH3 High Byte** = 0x00 default Reg. 0x2E **TACH4 Low Byte** = 0x00 default Reg. 0x2F **TACH4 High Byte** = 0x00 default

Reading Fan Speed from the ADT7460

If fan speeds are being measured, this involves a 2-register read for each measurement. The low byte should be read first. This causes the high byte to be frozen until both High and Low Byte registers have been read from. This prevents erroneous TACH readings.

The Fan Tachometer Reading registers report back the number of 11.11 µs period clocks (90 kHz oscillator) gated to the fan speed counter, from the rising edge of the first fan TACH pulse to the rising edge of the third fan TACH pulse (assuming two pulses per revolution are being counted). Since the device is essentially measuring the fan TACH period, the higher the count value the slower the fan is actually running. A 16-bit Fan Tachometer reading of 0xFFFF indicates either that the fan has stalled or is running very slowly (< 100 RPM).

HIGH LIMIT: > COMPARISON PERFORMED

Since the actual fan TACH period is being measured, exceeding a Fan TACH Limit by 1 will set the appropriate Status bit and can be used to generate an *SMBALERT*.

Fan TACH Limit Registers

The Fan TACH Limit Registers are 16-bit values consisting of two bytes.

Reg. 0x54 **TACH1 Minimum Low Byte** = 0xFF default

Reg. 0x55 **TACH1 Minimum High Byte** = 0xFF default

Reg. 0x56 **TACH2 Minimum Low Byte** = 0xFF default

Reg. 0x57 **TACH2 Minimum High Byte** = 0xFF default

Reg. 0x58 **TACH3 Minimum Low Byte** = 0xFF default

Reg. 0x59 **TACH3 Minimum High Byte** = 0xFF default

Reg. 0x5A **TACH4 Minimum Low Byte** = 0xFF default

Reg. 0x5B **TACH4 Minimum High Byte** = 0xFF default

Fan Speed Measurement Rate

The Fan TACH readings are normally updated once every second.

The FAST bit (Bit 3) of Configuration Register 3 (Reg. 0x78), when set, updates the Fan TACH readings every 250 ms.

If any of the fans are not being driven by a PWM channel but are powered directly from 5 V or 12 V, its associated dc bit in Configuration Register 3 should be set. This allows TACH readings to be taken on a continuous basis for fans connected directly to a dc source.

Calculating Fan Speed

Assuming a fan with a two pulses/revolution (and two pulses/rev being measured) fan speed is calculated by:

Fan Speed (RPM) = (90,000 60)/Fan Tach Reading

where:

Fan Tach Reading = 16-bit Fan Tachometer Reading

Example:

TACH1 High Byte (Reg $0x29$) = $0x17$ TACH1 Low Byte (Reg $0x28$) = $0xFF$

What is Fan 1 speed in RPM?

Fan 1 TACH reading $= 0x17FF = 6143$ decimal $RPM = (f \times 60)/Fan 1 TACH reading$ $RPM = (90000 \times 60)/6143$

Fan Speed = 879 RPM

Fan Pulses Per Revolution

Different fan models can output either 1, 2, 3, or 4 TACH pulses per revolution. Once the number of fan TACH pulses has been determined, it can be programmed into the Fan Pulses Per Revolution Register (Reg. 0x7B) for each fan. Alternatively, this register can be used to determine the number or pulses/revolution output by a given fan. By plotting fan speed measurements at 100% speed with different pulses/rev setting, the smoothest graph with the lowest ripple determines the correct pulses/rev value.

Fan Pulses Per Revolution Register

- \langle 1:0> FAN1 default = 2 pulses per rev.
- <3:2> FAN2 default = 2 pulses per rev.
- <5:4> FAN3 default = 2 pulses per rev.
- <7:6> FAN4 default = 2 pulses per rev.
	- $00 = 1$ pulse per rev. $01 = 2$ pulses per rev.
	- $10 = 3$ pulses per rev.
	-
	- $11 = 4$ pulses per rev.

2-Wire Fan Speed Measurements

The ADT7460 is capable of measuring the speed of 2-wire fans, i.e., fans without TACH outputs. To do this, the fan must be interfaced as shown in the Fan Drive Circuitry section of the data sheet. In this case, the TACH inputs need to be reprogrammed as analog inputs, AIN.

CONFIGURATION REGISTER 2 (REG. 0x73)

Bit 3 (AIN4) = 1, Pin 9 is reconfigured to measure the speed of a 2-wire fan using an external sensing resistor and coupling capacitor.

Bit 2 (AIN3) = 1, Pin 4 is reconfigured to measure the speed of a 2-wire fan using an external sensing resistor and coupling capacitor.

Bit 1 (AIN2) = 1, Pin 7 is reconfigured to measure the speed of a 2-wire fan using an external sensing resistor and coupling capacitor.

Bit 0 (AIN1) = 1, Pin 6 is reconfigured to measure the speed of a 2-wire fan using an external sensing resistor and coupling capacitor.

AIN Switching Threshold

Having configured the TACH inputs as AIN inputs for 2-wire measurements, you can select the sensing threshold for the AIN signal.

CONFIGURATION REGISTER 4 (REG. 0x7D)

<3:2> AINL These two bits define the input threshold for 2-wire fan speed measurements.

> $00 = \pm 20$ mV $01 = \pm 40$ mV $10 = \pm 80$ mV $11 = \pm 130$ mV

Fan Spin-Up

The ADT7460 has a unique fan spin-up function. It will spin the fan at 100% PWM duty cycle until two TACH pulses are detected on the TACH input. Once two pulses have been detected, the PWM duty cycle will go to the expected running value, e.g., 33%. The advantage of this is that fans have different spin-up characteristics and will take different times to overcome inertia. The ADT7460 just runs the fans fast enough to overcome inertia and will be quieter on spin-up than fans programmed to spin-up for a given spin-up time.

Fan Start-Up Timeout

To prevent false interrupts being generated as a fan spins up (since it is below running speed), the ADT7460 includes a Fan Start-Up Timeout function. This is the time limit allowed for two TACH pulses to be detected on spin-up. For example, if 2 seconds Fan Start-Up Timeout is chosen and no TACH pulses occur within 2 seconds of the start of spin-up, a fan fault is detected and flagged in the Interrupt Status Registers.

PWM1 CONFIGURATION (REG. 0x5C)

- **<2:0> SPIN** These bits control the Start-Up timeout for PWM1. **000 = No startup timeout 001 = 100 ms 010 = 250 ms (default) 011 = 400 ms 100 = 667 ms**
	- $101 = 1 s$ **110 = 2 s**
	- $111 = 4 s$

PWM2 CONFIGURATION (REG. 0x5D)

<2:0> SPIN These bits control the Start-Up timeout for PWM2. **000 = No startup timeout 001 = 100 ms 010 = 250 ms (default) 011 = 400 ms 100 = 667 ms** $101 = 1 s$ **110 = 2 s** $111 = 4 s$

PWM3 CONFIGURATION (REG. 0x5E)

<2:0> SPIN These bits control the Start-Up timeout for PWM3. **000 = No startup timeout 001 = 100 ms 010 = 250 ms (default) 011 = 400 ms 100 = 667 ms** $101 = 1 s$ **110 = 2 s** $111 = 4 s$

Disabling Fan Start-Up Timeout

Although Fan Start-Up makes fan spin-ups much quieter than fixed-time spin-ups, the option exists to use fixed spin-up times. Bit 5 (FSPDIS) = 1 in Configuration Register 1 (Reg. 0x40) disables the spin-up for two TACH pulses. Instead, the fan will spin up for the fixed time as selected in Registers 0x5C–0x5E.

PWM Logic State

The PWM outputs can be programmed high for 100% duty cycle (noninverted) or low for 100% duty cycle (inverted).

PWM2 Configuration (Reg. 0x5D) $\langle 4 \rangle$ **INV** $0 = \text{logic high for 100\%}$ PWM duty cycle $1 =$ logic low for 100% PWM duty cycle

PWM3 Configuration (Reg. 0x5E)

 $\langle 4 \rangle$ **INV** $0 = \text{logic high for } 100\%$ PWM duty cycle $1 =$ logic low for 100% PWM duty cycle

PWM Drive Frequency

The PWM drive frequency can be adjusted for the application. Registers 0x5F–0x61 configure the PWM frequency for PWM1–PWM3, respectively.

PWM1 FREQUENCY REGISTERS (REG. 0x5F–0x61)

Fan Speed Control

The ADT7460 can control fan speed using two different modes. The first is Automatic Fan Speed Control Mode. In this mode fan speed is automatically varied with temperature and without CPU intervention, once initial parameters are set up. The advantage of this is in the case of the system hanging, the user is guaranteed that the system is protected from overheating. The Automatic Fan Speed Control incorporates a feature called Dynamic T_min calibration. This feature reduces the design effort required to program the Automatic Fan Speed Control Loop. For more information and how to program the Automatic Fan Speed Control Loop and Dynamic T_min calibration; see the Automatic Fan Speed Control Loop application note.

The second fan speed control method is Manual Fan Speed Control which is described in the next paragraph.

Manual Fan Speed Control

The ADT7460 allows the Duty Cycle of any PWM output to be manually adjusted. This can be useful if you wish to change fan speed in software or want to adjust PWM duty cycle output for test purposes. Bits <7:5> of Registers 0x5C–0x5E (PWM Configuration) control the behavior of each PWM output.

PWM CONFIGURATION (REG. 0x5C–0x5E) $\langle 7:5 \rangle$ BHVR $111 =$ Manual Mode

Once under Manual Control, each PWM output may be manually updated by writing to registers 0x30–0x32 (PWMx Current Duty Cycle Registers).

Programming the PWMCurrent Duty Cycle Registers

The PWM Current Duty Cycle Registers are 8-bit registers that allow the PWM duty cycle for each output to be set anywhere from 0% (0x00) to 100% (0xFF) in steps of 0.39% (256 steps).

The value to be programmed into the PWM_{MIN} register is given by:

$$
Value(decimal) = PWM_{MIN} / 0.39
$$

Example 1: For a PWM duty cycle of 50%, *Value (decimal)* = 50/0.39 = 128 decimal Value = 128 decimal or $0x80$.

Example 2: For a PWM duty cycle of 33%,

Value (decimal) = 33/0.39 = 85 decimal

Value = 85 decimal or $0x54$.

PWM DUTY CYCLE REGISTERS

Reg. 0x30 PWM1 Duty Cycle = $0xFF(100\%$ default) Reg. 0x31 PWM2 Duty Cycle = $0xFF(100\%$ default) Reg. 0x32 PWM3 Duty Cycle = $0xFF(100\%$ default)

By reading the PWMx Current Duty Cycle Registers, you can keep track of the current duty cycle on each PWM output, even when the fans are running in Automatic Fan Speed Control Mode or Acoustic Enhancement Mode.

Figure 37. Control PWM Duty Cycle Manually with a Resolution of 0.39%

OPERATING FROM 3.3 V STANDBY

The ADT7460 has been specifically designed to operate from a 3.3 V STBY supply. In computers that support S3 and S5 states, the core voltage of the processor will be lowered in these states. If using the Dynamic T_{MIN} Mode, lowering the core voltage of the processor would change the CPU temperature and change the dynamics of the system under dynamic $T_{\rm MIN}$ control. Likewise, when monitoring *THERM*, the *THERM* timer should be disabled during these states.

XOR TREE TEST MODE

The ADT7460 includes an XOR Tree Test Mode. This mode is useful for in-circuit test equipment at board-level testing. By applying stimulus to the pins included in the XOR Tree, it is possible to detect opens or shorts on the system board. Figure 38 shows the signals that are exercised in the XOR Tree Test Mode.

Figure 38. XOR Tree Test

The XOR Tree Test is invoked by setting Bit 0 (XEN) of the XOR Tree Test Enable Register (Reg. 0x6F).

Table IV. ADT7460 Registers (continued)

These voltage readings are in twos complement format.

If the extended resolution bits of these readings are also being read, the Extended Resolution registers (Reg. 0x76, 0x77) should be read first. Once the Extended Resolution registers get read, the associated MSB reading registers get frozen until read. Both the Extended Resolution Registers and the MSB registers get frozen.

These temperature readings are in twos complement format.

*Note that a reading of 0x80 in a temperature reading register indicates a diode fault (open or short) on that channel. If the extended resolution bits of these readings are also being read, the Extended Resolution registers (Reg. 0x76, 0x77) should be read first. Once the Extended Resolution registers get read, all associated MSB reading registers get frozen until read. Both the Extended Resolution Registers and the MSB registers get frozen.

These registers count the number of 11.11 µs periods (based on an internal 90 kHz clock) that occur between a number of consecutive fan TACH pulses (default = 2). The number of TACH pulses used to count can be changed using the Fan Pulses Per Revolution Register (Reg. 0x7B). This allows the fan speed to be accurately measured. Since a valid Fan Tachometer reading requires that two bytes are read, the low byte MUST be read first. Both the low and high bytes are then frozen until read. At power-on, these registers contain 0x0000 until such time as the first valid fan TACH measurement is read in to these registers. This prevents false interrupts from occurring while the fans are spinning up.

- A count of 0xFFFF indicates that a fan is: **1. Stalled or Blocked** (object jamming the fan)
	- **2. Failed** (internal circuitry destroyed)
	- **3. Not Populated** (The ADT7460 expects to see a fan connected to each TACH. If a fan is not connected to that TACH, its TACH minimum high and low byte should be set to 0xFFFF.)
	- **4. Alternate Function**, e.g., TACH4 reconfigured as *THERM* pin
	- **5. 2-Wire Instead of 3-Wire Fan**

Table VIII. Current PWM Duty Cycle Registers (Power-On Default = 0xFF)

These registers reflect the PWM duty cycle driving each fan at any given time. When in Automatic Fan Speed Control Mode, the ADT7460 reports the PWM duty cycles back through these registers. The PWM duty cycle values will vary according to temperature in Automatic Fan Speed Control Mode. During fan startup, these registers report back 0x00. In Software Mode, the PWM duty cycle outputs can be set to any duty cycle value by writing to these registers.

Table IX. Operating Point Registers (Power-On Default = 0x64)

These registers set the target Operating Point for each temperature channel when the Dynamic T_{MIN} Control feature is enabled.

The fans being controlled will be adjusted to maintain temperature about an Operating Point.

*These registers become read-only when the Configuration Register 1 Lock bit is set to 1. Any subsequent attempts to write to these registers will fail.

Table X. Register 0x36 - Dynamic T_{MIN} Control Register 1 (Power-On Default = 0x00)

*This register becomes read-only when the Configuration Register 1 Lock bit is set to 1. Any subsequent attempts to write to this register will fail.

*This register becomes read-only when the Configuration Register 1 Lock bit is set to 1. Any subsequent attempts to write to this register will fail.

Table XII. Register 0x40 – Configuration Register 1 (Power-On Default = 0x00)

Table XIII. Register 0x41 – Interrupt Status Register 1 (Power-On Default = 0x00)

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Table XIV. Register 0x42 – Interrupt Status Register 2 (Power-On Default = 0x00)

Table XV. Voltage Limit Registers

Setting the Configuration Register 1 Lock bit has no effect on these registers.

High Limits: An interrupt is generated when a value exceeds its high limit (> comparison).

Low Limits: An interrupt is generated when a value is equal to or below its low limit (≤ comparison).

Table XVI. Temperature Limit Registers

Exceeding any of these temperature limits by 1°C will cause the appropriate status bit to be set in the Interrupt Status Register. Setting the Configuration Register 1 Lock bit has no effect on these registers.

High Limits: An interrupt is generated when a value exceeds its high limit (> comparison).

Low Limits: An interrupt is generated when a value is equal to or below its low limit (\leq comparison).

Table XVII. Fan Tachometer Limit Registers

Exceeding any of the TACH limit registers by 1 indicates that the fan is running too slowly or has stalled. The appropriate status bit will be set in Interrupt Status Register 2 to indicate the fan failure. Setting the Configuration Register 1 Lock bit has no effect on these registers.

Table XVIII. PWM Configuration Registers

*These registers become read-only when the Configuration Register 1 Lock bit is set to 1. Any subsequent attempts to write to these registers will fail.

Table XIX. TEMP T_{RANGE}/PWM Frequency Registers

*These registers become read-only when the Configuration Register 1 Lock bit is set. Any further attempts to write to these registers shall have no effect.

Table XX. Register 0x62 – Enhance Acoustics Register 1 (Power-On Default = 0x00)

*This register becomes read-only when the Configuration Register 1 Lock bit is set to 1. Any further attempts to write to this register will have no effect.

Table XXI. Register 0x63 – Enhance Acoustics Register 2 (Power-On Default = 0x00)

*This register becomes read-only when the Configuration Register 1 Lock bit is set to 1. Any further attempts to write to this register will have no effect.

Table XXII. PWM Min Duty Cycle Registers

*These registers become read-only when the ADT7460 is in Automatic Fan Control Mode.

Table XXIII. T_{MIN} Registers

These are the T_{MIN} registers for each temperature channel. When the temperature measured exceeds T_{MIN}, the appropriate fan will run at minimum speed and increase with temperature according to $T_{\rm{RANGE}}$.

*These registers become read-only when the Configuration Register 1 Lock bit is set. Any further attempts to write to these registers shall have no effect.

Table XXIV. THERM Limit Registers

If any temperature measured exceeds its *THERM* limit, all PWM outputs will drive their fans at 100% duty cycle. This is a fail-safe mechanism incorporated to cool the system in the event of a critical overtemperature. It also ensures some level of cooling in the event that software or hardware locks up. If set to 0x80, this feature is disabled. The PWM output will remain at 100% until the temperature drops below *THERM* limit – Hysteresis. If the *THERM* pin is programmed as an output, then exceeding these limits by 0.25° C can cause the \overline{THERM} pin to assert low as an output.

*These registers become read-only when the Configuration Register 1 Lock bit is set to 1. Any further attempts to write to these registers will have no effect.

Table XXV. Temperature Hysteresis Registers

Each 4-bit value controls the amount of temperature hysteresis applied to a particular temperature channel. Once the temperature for that channel falls below its T_{MIN} value, the fan will remain running at PWM_{MIN} duty cycle until the temperature = T_{MIN} – Hysteresis. Up to 15°C of hysteresis may be assigned to any temperature channel. The hysteresis value chosen will also apply to that temperature channel if its *THERM* limit is exceeded. The PWM output being controlled will go to 100% if the *THERM* limit is exceeded and will remain at 100% until the temperature drops below *THERM* – Hysteresis. For acoustic reasons, it is recommended that the hysteresis value not be programmed less than 4° C. Setting the hysteresis value lower than 4° C will cause the fan to switch on and off regularly when the temperature is close to T_{MIN} .

*These registers become read-only when the Configuration Register 1 Lock bit is set to 1. Any further attempts to write to these registers will have no effect.

Table XXVI. XOR Tree Test Enable

*This register becomes read-only when the Configuration Register 1 Lock bit is set to 1. Any further attempts to write to this register will have no effect.

Table XXVII. Remote 1 Temperature Offset

*This register becomes read-only when the Configuration Register 1 Lock bit is set to 1. Any further attempts to write to this register will have no effect.

Table XXVIII. Local Temperature Offset

*This register becomes read-only when the Configuration Register 1 Lock bit is set to 1. Any further attempts to write to this register will have no effect.

Table XXIX. Remote 2 Temperature Offset

*This register becomes read-only when the Configuration Register 1 Lock bit is set to 1. Any further attempts to write to this register will have no effect.

Table XXX. Register 0x73 – Configuration Register 2 (Power-On Default = 0x00)

*This register becomes read-only when the Configuration Register 1 Lock bit is set to 1. Any further attempts to write to this register will have no effect.

Table XXXII. Register 0x75 – Interrupt Mask Register 2 (Power On Default <7:0> = 0x00)

Table XXXIII. Register 0x76 – Extended Resolution Register 1

If this register is read, this register and the registers holding the MSB of each reading are frozen until read.

Table XXXIV. Register 0x77 – Extended Resolution Register 2

If this register is read, this register and the registers holding the MSB of each reading are frozen until read.

Table XXXV. Register 0x78 – Configuration Register 3 (Power-On Default = 0x00)

*This register becomes read-only when the Configuration Register 1 Lock bit is set to 1. Any further attempts to write to this register will have no effect.

Table XXXVI. Register 0x79 – THERM Status Register (Power-On Default = 0x00)

Table XXXVII. Register 0x7A – THERM Limit Register (Power-On Default = 0x00)

Table XXXVIII. Register 0x7B – Fan Pulses Per Revolution Register (Power On Default = 0x55)

Table XXXIX. REGISTER 0x7D – Configuration Register 4 (Power-On Default = 0x00)

*This register becomes read-only when the Configuration Register 1 Lock bit is set to 1. Any further attempts to write to this register will have no effect.

Table XL. Register 0x7E – Manufacturer's Test Register 1 (Power On-Default = 0x00)

Table XLI. Register 0x7F – Manufacturer's Test Register 2 (Power-On Default = 0x00)

OUTLINE DIMENSIONS

16-Lead SOIC, 0.025 Lead Pitch [QSOP] (RQ-16)

Dimensions shown in millimeters

COMPLIANT TO JEDEC STANDARDS MO-137AB

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